

## Chapter 5, The DTFT (Discrete-Time Fourier Transform): Problem Solutions

### Problem 1

1. The inner product in  $l_2(\mathbb{Z})$  is defined as

$$\langle x[n], y[n] \rangle = \sum_n x^*[n]y[n],$$

and in  $L_2([-\pi, \pi])$  as

$$\langle X(e^{jw}), Y(e^{jw}) \rangle = \int_{-\pi}^{\pi} X^*(e^{jw})Y(e^{jw})dw.$$

Thus,

$$\begin{aligned} \frac{1}{2\pi} \int_{-\pi}^{\pi} X^*(e^{jw})Y(e^{jw})dw &= \frac{1}{2\pi} \int_{-\pi}^{\pi} (\sum_n x[n]e^{-jwn})^* \sum_m y[m]e^{-jwm}dw \\ &\stackrel{(1)}{=} \frac{1}{2\pi} \int_{-\pi}^{\pi} \sum_n x^*[n]e^{jwn} \sum_m y[m]e^{-jwm}dw \\ &= \frac{1}{2\pi} \int_{-\pi}^{\pi} \sum_n \sum_m x^*[n]y[m]e^{jw(n-m)}dw \\ &\stackrel{(2)}{=} \frac{1}{2\pi} \sum_n \sum_m x^*[n]y[m] \int_{-\pi}^{\pi} e^{jw(n-m)}dw \\ &\stackrel{(3)}{=} \sum_n x^*[n]y[n], \end{aligned}$$

where (1) follows from the properties of the complex conjugate, (2) follows from swapping the integral and the sums and (3) from the fact that

$$\frac{1}{2\pi} \int_{-\pi}^{\pi} e^{jw(n-m)}dw = \begin{cases} 1 & \text{if } m = n \\ 0 & \text{if } m \neq n. \end{cases}$$

2. If  $x[n] = y[n]$ , then  $\langle x[n], x[n] \rangle$  corresponds to the energy of the signal in the time domain and  $\langle X(e^{jw}), X(e^{jw}) \rangle$  to the energy of the signal in the frequency domain. In this case, the Plancherel-Parseval equality illustrates an energy conservation property from the time domain to the frequency domain. This property is known as the *Parseval theorem*.

### Problem 2

[DFT AND DTFT]

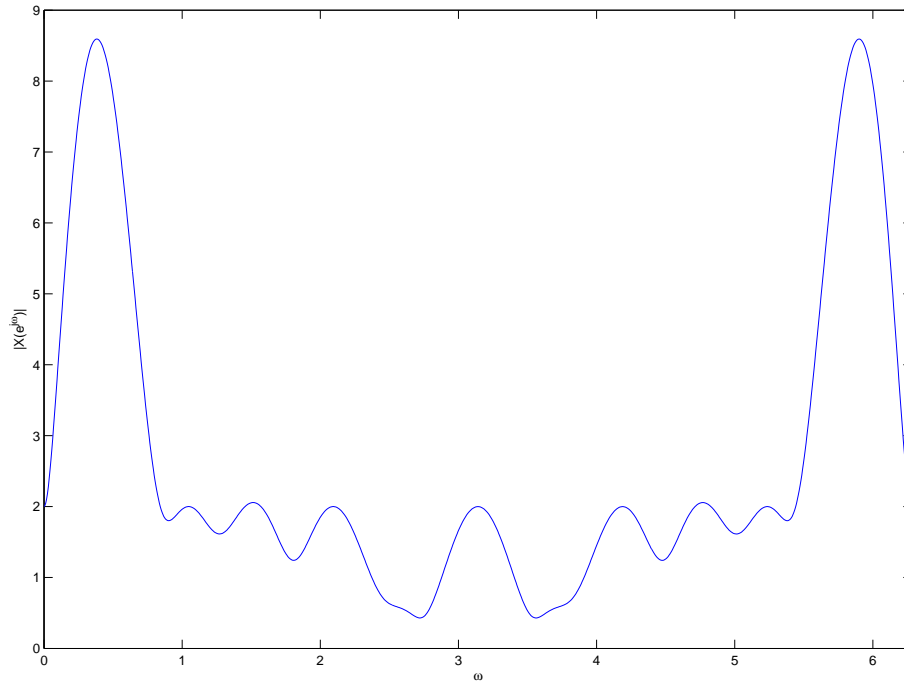


Figure 1: Problem 5(a).

(a)

$$\begin{aligned}
 X(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n} \\
 &= \sum_{n=3}^9 e^{-j\omega n} - \sum_{n=10}^{14} e^{-j\omega n} \\
 &= e^{-j\omega 3} \frac{1 - e^{-j\omega 7}}{1 - e^{-j\omega}} - e^{-j\omega 10} \frac{1 - e^{-j\omega 5}}{1 - e^{-j\omega}} \\
 &= \frac{e^{-j\omega 3} - 2e^{-j\omega 10} + e^{-j\omega 15}}{1 - e^{-j\omega}},
 \end{aligned}$$

for  $\omega \neq 0$ . For  $\omega = 0$  we have  $X(e^{j0}) = \sum_{n=3}^9 1 - \sum_{n=10}^{14} 1 = 2$ .

(b) See Figure 1 that was obtained by

```

>> w = linspace(0,2*pi,1e3+1);
>> w = w(1:end-1); % we don't want 2*pi itself
>> X = (exp(-i*w*3)-2*exp(-i*w*10)+exp(-i*w*15))./(1-exp(-i*w));
Warning: Divide by zero.
>> X(1) = 2;
>> plot(w,abs(X))
>> xlim([0 2*pi])
>> xlabel('\omega')
>> ylabel('|X(e^{j\omega})|')

```

(c)-(d) See Figure 2, where we give the result for  $N = 100$  only:

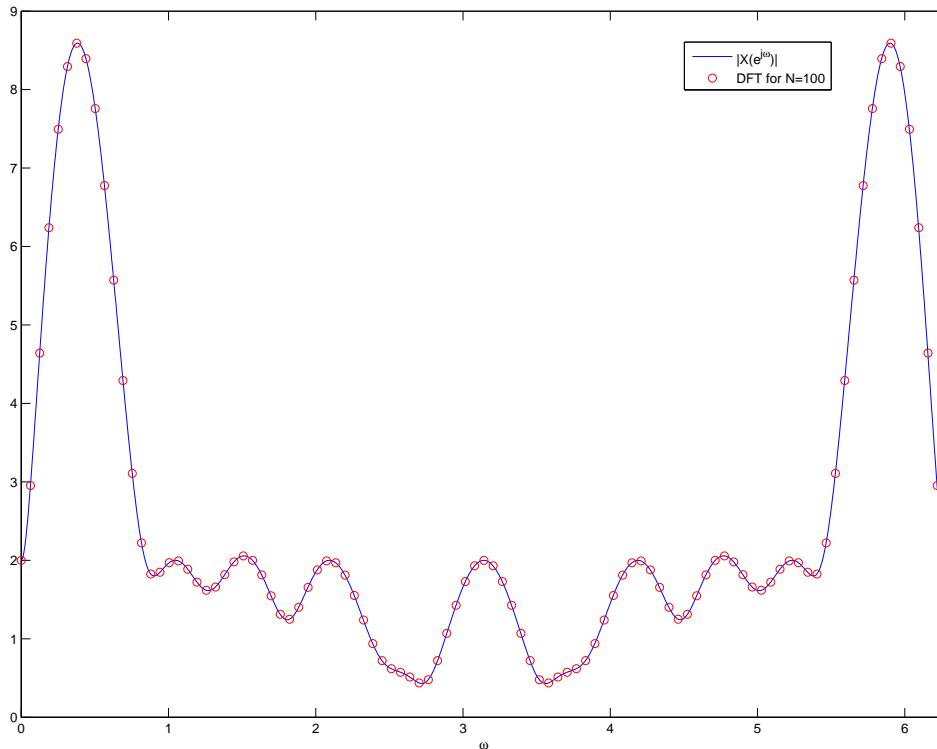


Figure 2: Problem 6(c)-(d).

```

>> N=1e2;
>> Xe2 = myDFT(x,N);
>> plot(w,abs(X))
>> xlim([0 2*pi])
>> hold on
>> plot([0:N-1]/N*2*pi,abs(Xe2),'or');
>> xlabel('\omega')
>> legend('|X(e^{j\omega})|','DFT for N=100');

```

We see that the the DFT sequence corresponds exactly to points on the DTFT curve.

### Problem 3

1. The discrete-time sequence  $x[n]$  can be written as the convolution of  $x_1[n]$  and  $x_2[n]$  defined as

$$x_1[n] = x_2[n] = \begin{cases} 1 & -(M-1)/2 \leq n \leq (M-1)/2 \\ 0 & \text{otherwise.} \end{cases}$$

In fact,

$$\begin{aligned}
 x_1[n] * x_2[n] &= \sum_k x_1[k] x_2[n-k] \\
 &\stackrel{(1)}{=} \sum_k x_1[k] x_1[k-n] \\
 &\stackrel{(2)}{=} x[n]
 \end{aligned}$$

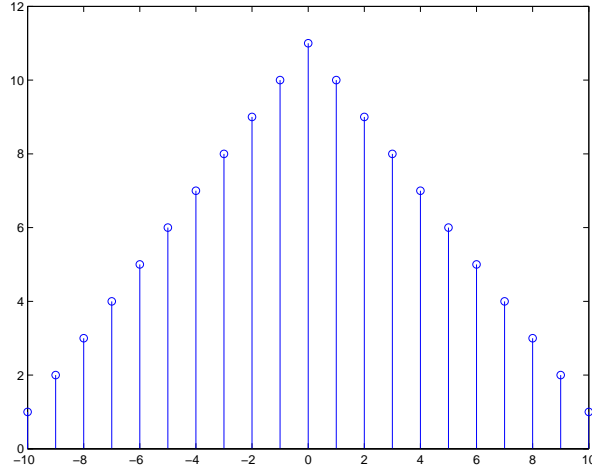


Figure 3: The discrete-time sequence  $x[n]$  for  $M = 11$ .

where (1) follows from the fact that  $x_1[n] = x_2[n]$  and from the symmetry of  $x_1[n]$  and (2) noticing that the sum corresponds to the size of the overlapping area between  $x_1[k]$  and its  $n$ -shifted version  $x_1[k - n]$ . When  $|n| \geq M$  the two sequences do not overlap whereas the size of the overlapping area reaches its maximum  $M$  when  $n = 0$ .

Using Matlab, we can easily verify the above result for  $M = 11$  using the following code:

```
>> M = 11;
>> x1 = ones(1,M);
>> x2 = x1;
>> x = conv(x1,x2);
>> stem([-M+1:M-1], x);
```

The result is shown in Figure 3.

- Note that  $x_1[n] = u[n + (M - 1)/2] - u[n - (M + 1)/2]$ . We can thus compute its DTFT as

$$\begin{aligned}
 X_1(e^{j\omega}) &\stackrel{(1)}{=} \left( \frac{1}{1 - e^{-j\omega}} + \frac{1}{2} \tilde{\delta}(\omega) \right) \left( e^{j\omega(M-1)/2} - e^{-j\omega(M+1)/2} \right) \\
 &\stackrel{(2)}{=} \frac{e^{j\omega(M-1)/2} - e^{-j\omega(M+1)/2}}{1 - e^{-j\omega}} = \frac{e^{-j\omega/2}(e^{j\omega M/2} - e^{-j\omega M/2})}{e^{-j\omega/2}(e^{j\omega/2} - e^{-j\omega/2})} \\
 &= \frac{\sin(\omega M/2)}{\sin(\omega/2)}
 \end{aligned}$$

where (1) follows from the DTFT of  $u[n]$  and (2) from the fact that

$$e^{j\omega(M-1)/2} \tilde{\delta}(\omega) = e^{-j\omega(M+1)/2} \tilde{\delta}(\omega) = \tilde{\delta}(\omega).$$

Using the convolution theorem, we can write

$$\begin{aligned}
 X(e^{j\omega}) &= X_1(e^{j\omega})X_2(e^{j\omega}) \\
 &= X_1(e^{j\omega})X_1(e^{j\omega}) \\
 &= \left( \frac{\sin(\omega M/2)}{\sin(\omega/2)} \right)^2.
 \end{aligned}$$

## Problem 4

1.  $\mathcal{H}\{\delta[n]\} = \delta[n]$ ; but  $\mathcal{H}\{a\delta[n]\} = a^2\delta[n] \neq a\mathcal{H}\{\delta[n]\}$ .
2. Let  $y[n] = \mathcal{H}\{x[n]\}$ ; let  $w[n] = x[n - n_0]$ ;  $\mathcal{H}\{w[n]\} = w^2[n] = x^2[n - n_0] = y[n - n_0]$ . QED.
3. First of all,  $y[n] = \cos^2(\omega_0 n) = (1 + \cos(2\omega_0 n))/2$  from the well-known trigonometric identity. So  $y[n]$  contains a sinusoid at *double* the original frequency (but be careful: double in the  $2\pi$ -periodic sense: if  $\omega_0$  is larger than  $\pi/2$ , then  $2\omega_0$  will wrap around the  $[-\pi, \pi]$  interval).  
If  $\omega_0 = 3\pi/8$ , then  $y[n] = (1 + \cos((3\pi/4)n))/2$ ; since  $\mathcal{G}$  is a highpass with cutoff frequency  $\pi/2$ , it will kill the frequency components below  $\pi/2$  and therefore it will kill the constant. The only component that passes through is the cosine at  $3\pi/4$ . The final output is therefore  $v[n] = \frac{1}{2} \cos((3\pi/4)n)$ .
4. If  $\omega_0 = 7\pi/8$ , then  $2\omega_0 = 7\pi/4 > \pi$ . We can therefore bring back the frequency into the  $[-\pi, \pi]$  interval. We have that  $7\pi/4 = 2\pi - \pi/4$  and therefore  $\cos((7\pi/4)n) = \cos((2\pi - \pi/4)n) = \cos((\pi/4)n)$ . So in the end  $y[n] = (1 + \cos((\pi/4)n))/2$ . Now the frequency of the cosine is below  $\pi/2$  and therefore  $v[n] = 1 + \cos((\pi/4)n)$ . Note that, as for most nonlinear systems, the frequency of the input sinusoid is different from the frequency of the output sinusoids: sinusoids are no longer eigenfunctions!